

Filtering (Digital Systems)



Convolution of Sequences

$$x[n] \quad h[n] \quad y[n] = h[n] * x[n]$$

Discrete Fourier Transform (DFT)

$$\tilde{X}[k] \quad \tilde{T}[k] \quad \tilde{Y}[k] = \tilde{T}[k]\tilde{X}[k]$$

z-Transform

$$X[z] \quad T[z] \quad Y[z] = T[z]X[z]$$

The z-Transform

$$Z\{x[n]\} = \sum_{n=-\infty}^{\infty} x[n]z^{-n} = X(z)$$

Properties

- $x_1[n]*x_2[n] = \sum_{m=-\infty}^{\infty} x_1[m]x_2[n-m] \Leftrightarrow X_1(z) \cdot X_2(z)$ (convolution theorem)
- $x[n - n_0] \Leftrightarrow z^{-n_0}X(z)$ (shifting theorem)
- $x[-n] \Leftrightarrow X(1/z)$

Transfer Function

$$T(z) = \frac{Z\{y[n]\}}{Z\{x[n]\}} = \frac{Y(z)}{X(z)}$$

Rational Transfer Function \Leftrightarrow **Linear Difference Equation**

$$T(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{l=0}^M b_l z^{-l}}{\sum_{k=0}^N a_k z^{-k}} \Leftrightarrow \sum_{k=0}^N a_k y[n-k] = \sum_{l=0}^M b_l x[n-l]$$

Recursive/Non-Recursive Filters

$$\sum_{k=0}^N a_k y[n-k] = \sum_{l=0}^M b_l x[n-l]$$

$$a_0 y[n] + \sum_{k=1}^N a_k y[n-k] = \sum_{l=0}^M b_l x[n-l]$$

$$y[n] = -\sum_{k=1}^N \frac{a_k}{a_0} y[n-k] + \sum_{l=0}^M \frac{b_l}{a_0} x[n-l]$$

recursive

non-recursive

(IIR Filters)

FIR Filters

$$(a_0 = 1 \text{ and } a_k = 0 \text{ for } k \geq 1)$$

$$y[n] = \sum_{l=0}^M b_l x[n-l]$$

Rational Transfer Function

$$T(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{l=0}^M b_l z^{-l}}{\sum_{k=0}^N a_k z^{-k}}$$

Special Case: FIR filter ($a_0 = 1$ and $a_k = 0$ for $k \geq 1$)

$$\begin{aligned} T(z) &= \sum_{l=0}^M b_l z^{-l} = b_0 \prod_{l=1}^M (1 - c_l z^{-1}) \\ &= z^{-M} \cdot b_0 \prod_{l=1}^M (z - c_l) \end{aligned}$$

Linear Difference Equation

$$y[n] = \sum_{l=0}^M b_l x[n-l] = x[n] * b[l]$$

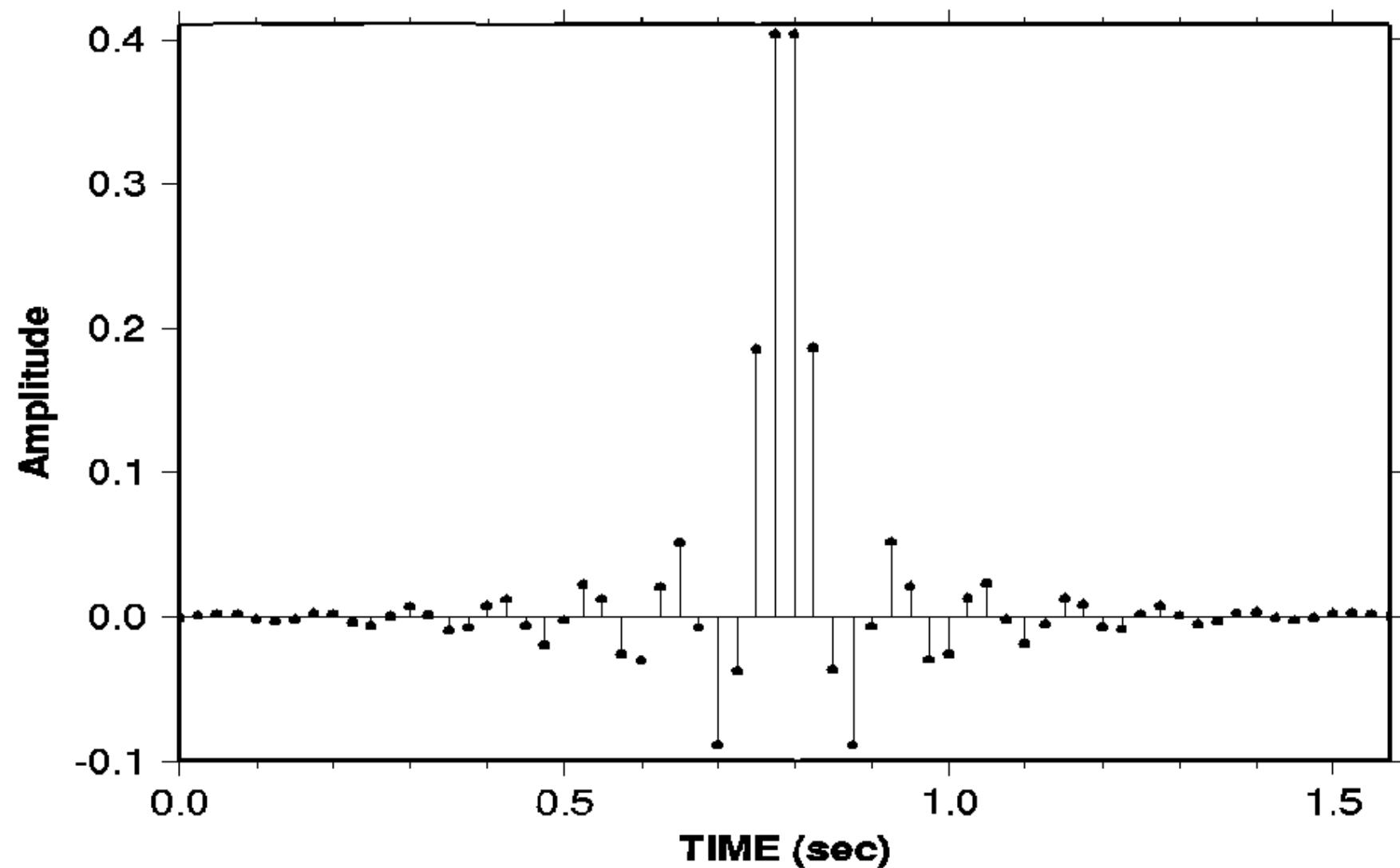
- FIR filters:

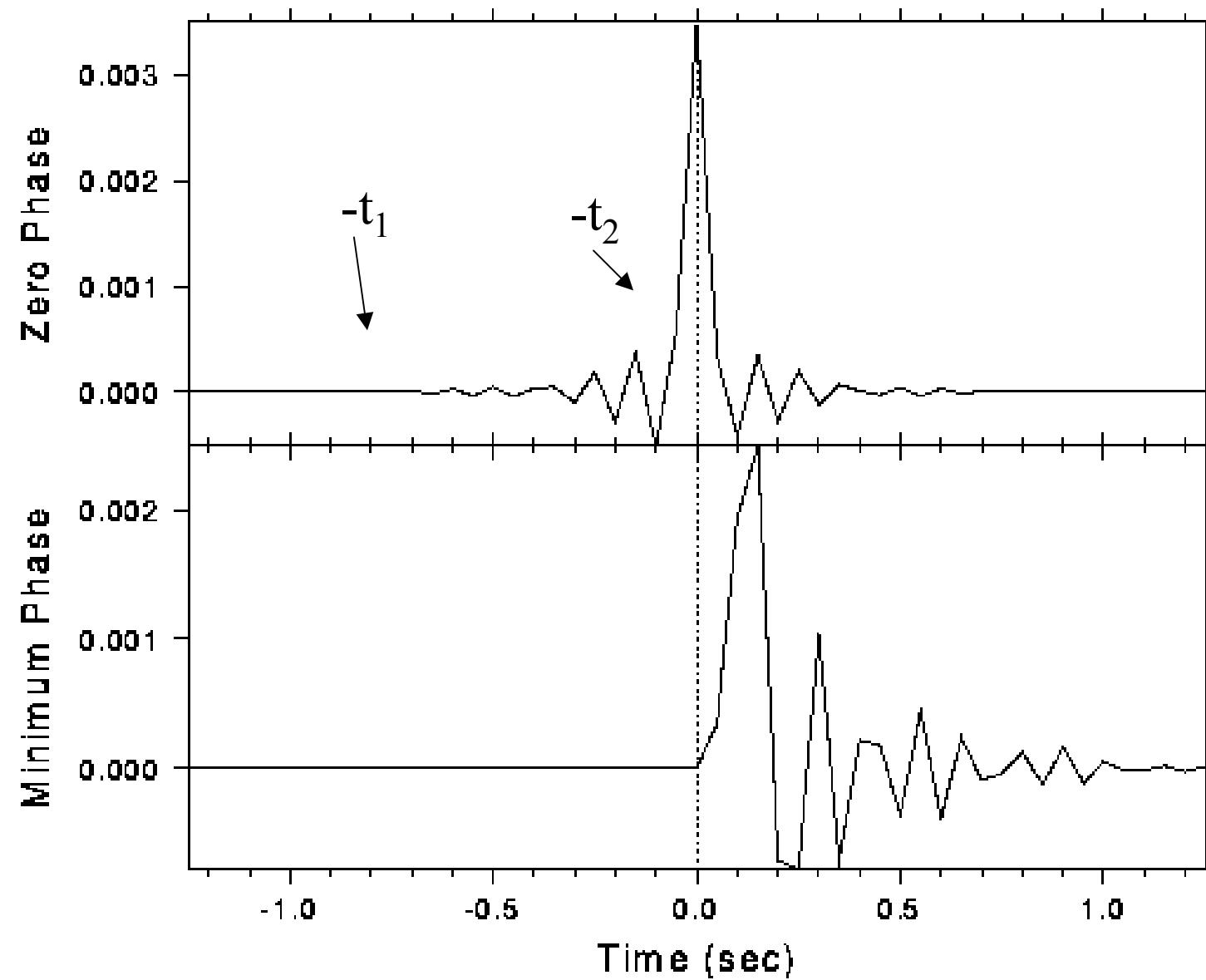
- + Always stable.
- Steep filters need many coefficients.
- + Both causal and noncausal filters can be implemented.
- + Filters with given specifications are easy to implement!

- IIR filters:

- Potentially unstable and subject to quantization errors.
- + Steep filters can easily be implemented with a few coefficients. Speed.
- Filters with given specifications are in general, difficult, if not impossible, to implement *exactly*(!).

QDP 380 Stage 4





**How can we
remove
these effects?**

Back to
Transfer function:

Recursive/Non-Recursive Filters

$$\sum_{k=0}^N a_k y[n-k] = \sum_{l=0}^M b_l x[n-l]$$

$$a_0 y[n] + \sum_{k=1}^N a_k y[n-k] = \sum_{l=0}^M b_l x[n-l]$$

$$y[n] = -\sum_{k=1}^N \frac{a_k}{a_0} y[n-k] + \sum_{l=0}^M \frac{b_l}{a_0} x[n-l]$$

recursive

non-recursive

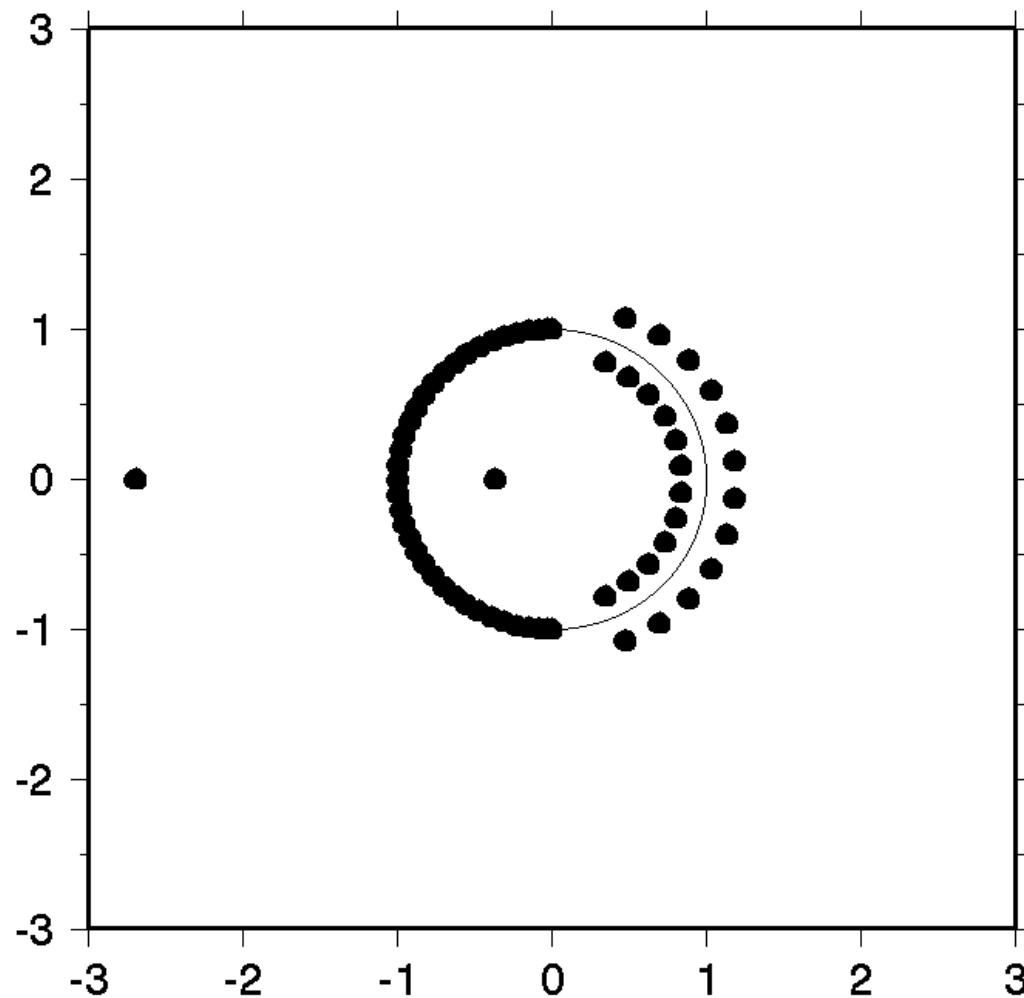
(IIR Filters)

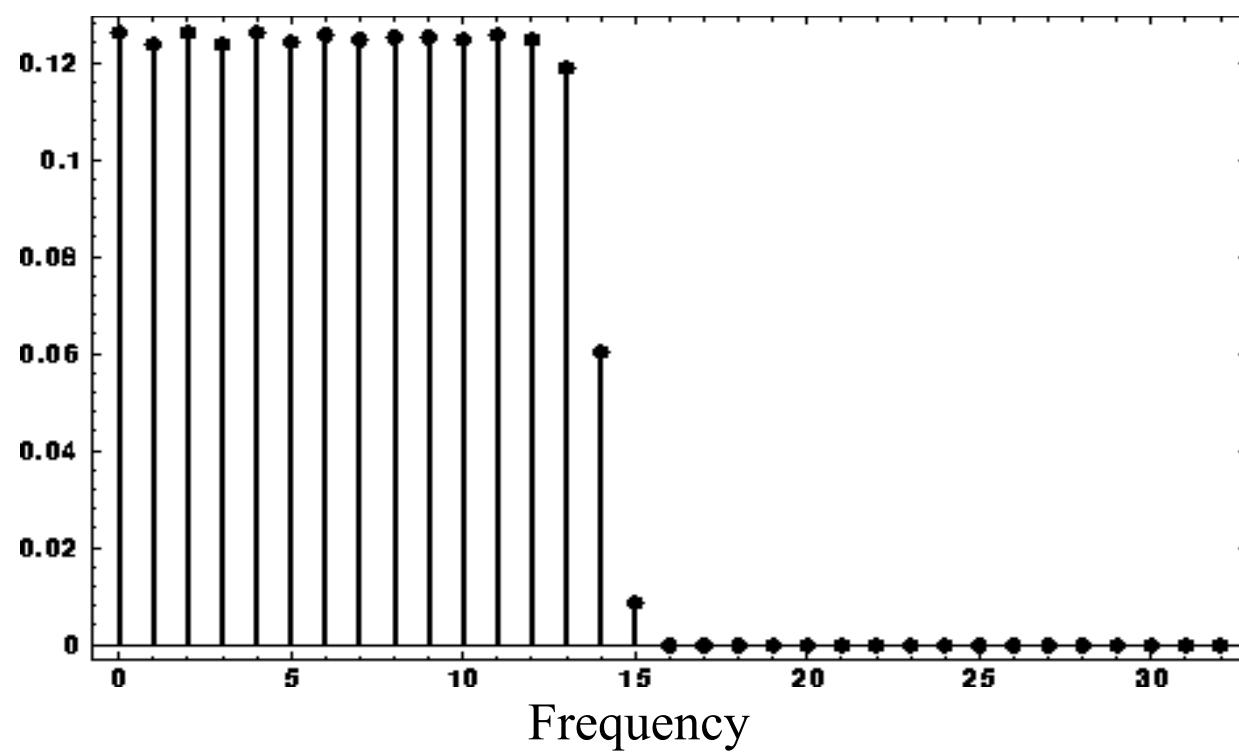
FIR Filters

$$(a_0 = 1 \text{ and } a_k = 0 \text{ for } k \geq 1)$$

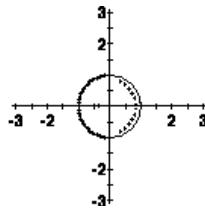
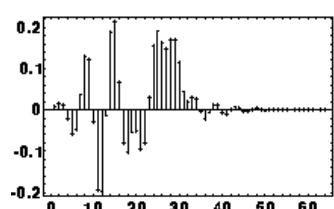
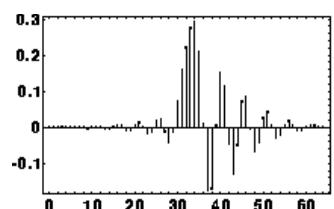
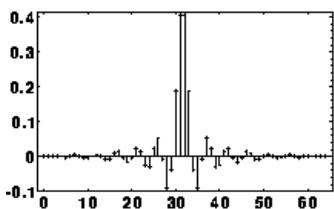
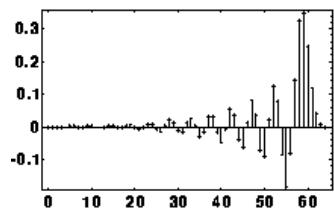
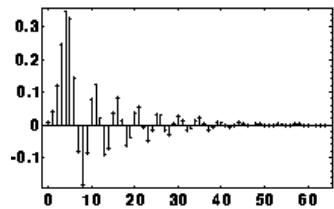
$$y[n] = \sum_{l=0}^M b_l x[n-l]$$

QDP 380 Stage 4

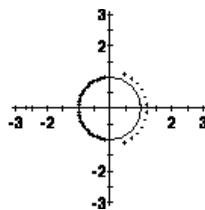
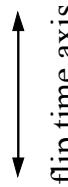




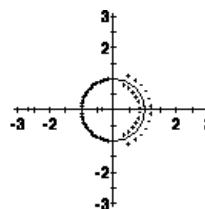
Waveform Properties and Root Positions



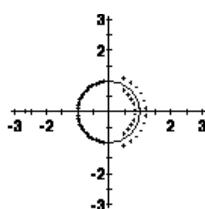
minimum delay/phase



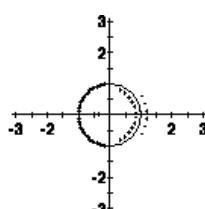
maximum delay/phase



mixed delay/phase
(zero phase)



mixed delay/phase



mixed delay/phase

Zero Phase FIR Filter

Problem: Two-Sided IR

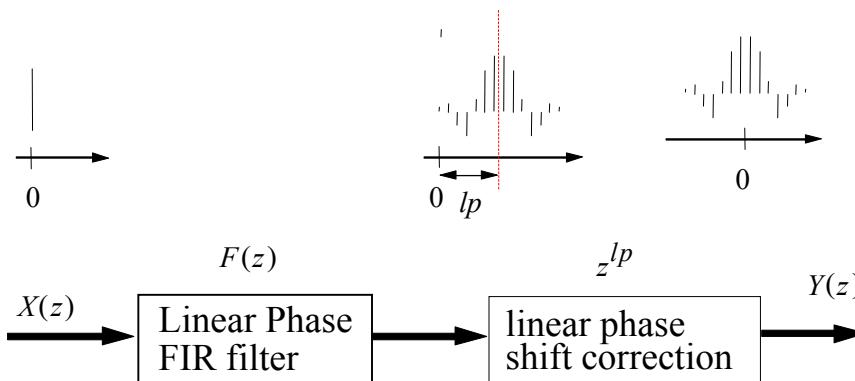
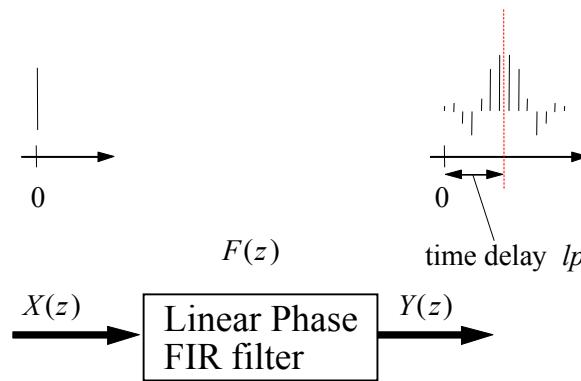
Cure: Change IR into Minimum Phase

Methods:

- 1) Add phase of Minimum Phase Filter to trace spectrum
- 2) Recursive Filtering of time inverted trace

Removing the acausal response of a Zero Phase FIR Filter

Linear Phase and Zero Phase Filter:



Linear-phase FIR Filter: $F(z)$

Zero-phase FIR Filter: $F(z) = F(z) \cdot z^{lp}$

z^{lp} = Time delay correction by lp samples

General rule:

Any filter can be expressed by convolution of its minimum phase and maximum phase component.

roots within UC

roots outside UC

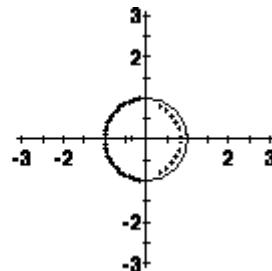
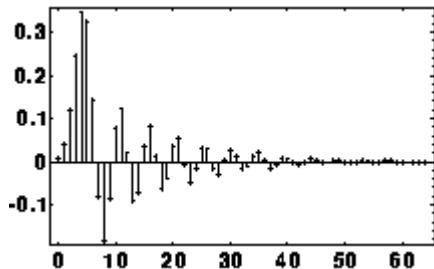
=> Linear phase FIR Filter: $F(z) = F_{max}(z) \cdot F_{min}(z)$

maximum phase component -> left-sided (acausal) component
minimum phase component -> right-sided (causal) component

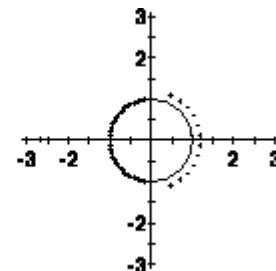
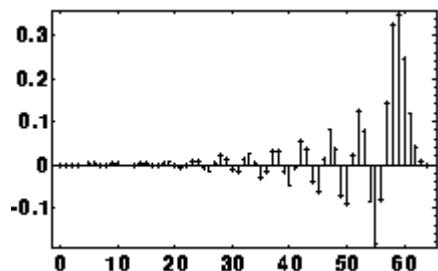
Removal of acausal response:

Replace maximum phase component $F_{max}(z)$ by its minimum phase equivalent $MinPhase\{F_{max}(z)\}$.

$$MinPhase\{F_{max}(z)\} =$$



minimum phase



maximum phase

Answer: flip maximum phase component in time.

In terms of z-transform:

$$\Rightarrow MinPhase\{F_{max}(z)\} = F_{max}(1/z)$$

z-transform representation of digital seismogram

$$\widetilde{Y}(z) = \underbrace{F(z) \cdot z^{lp}}_{\text{LP FIR}} \cdot \widetilde{X}(z) \underbrace{\cdot z^{-lp}}_{\text{LP correction}}$$

digital seismogram
after FIR filtering
but before decimation

digital seismogram
before FIR filtering

$\widetilde{x}[n]$ = unfiltered digital seismogram

$\widetilde{X}(z)$ = z-transform of the input signal $\widetilde{x}[n]$

$F(z)$ = z-transform of the linear phase FIR filter

$\widetilde{y}[n]$ = filtered digital seismic trace before decimation¹

$\widetilde{Y}(z)$ = z-transform of $\widetilde{y}[n]$

¹ This is a fictitious signal since decimation is commonly done while filtering.

$$F(z) \cdot z^{lp}$$

corresponds to a zero phase filter in which the linear phase component of $F(z)$ is corrected.

In practice: Treat time shift z^{lp} separately from $F(z)$

Removing the maximum phase component of a FIR filter

Principle:

$$F_{max}(z) \rightarrow MinPhase\{F_{max}(z)\}$$

‘Corrected’ seismogram $\tilde{Y}(z)$: $Y(z) = \frac{1}{F_{max}(z)} \cdot F_{max}(1/z) \cdot \tilde{Y}(z)$

Problem: Since $F_{max}(z)$ has only zeros outside the unit circle, $1/F_{max}(z)$ will have poles outside the unit circle.

Solution: Flip time axis.

$$Y(1/z) = \frac{1}{F_{max}(1/z)} \cdot F_{max}(z) \cdot \tilde{Y}(1/z)$$

Impulse response corresponding to $1/ F_{max}(z)$ becomes a stable causal sequence in nominal time and the deconvolution of the maximum phase component $F_{max}(z)$ poses no stability problems.

The difference equation

$$F_{max}(1/z) \cdot Y(1/z) = F_{max}(z) \cdot \tilde{Y}(1/z)$$

Rewrite to

$$A'(z) \cdot Y(z) = B'(z) \cdot X'(z)$$

$$A'(z) \Leftrightarrow F_{max}(1/z) \quad Y(z) \Leftrightarrow Y(1/z)$$

$$B'(z) \Leftrightarrow F_{max}(z) \quad X'(z) \Leftrightarrow \tilde{Y}(1/z)$$

Written as convolution sum:

$$\sum_{k=-\infty}^{\infty} a'[k] \cdot y'[i-k] = \sum_{l=-\infty}^{\infty} b'[l] \cdot x'[i-l]$$

Assumption: $F(z)$ contains mx zeros outside the unit circle
 \Rightarrow wavelets $a'[k]$ and $b'[k]$ will be of length $mx + 1$.

$$\sum_{k=0}^{mx} a'[k] \cdot y'[i-k] = \sum_{l=0}^{mx} b'[l] \cdot x'[i-l]$$

Rearrange to

$$y'[i] \cdot a'[0] + \sum_{k=1}^{mx} a'[k] \cdot y'[i-k] = \sum_{l=0}^{mx} b'[l] \cdot x'[i-l]$$

which is equivalent to

$$y'[i] = - \sum_{k=1}^{mx} \frac{a'[k]}{a'[0]} \cdot y'[i-k] + \sum_{l=0}^{mx} \frac{b'[l]}{a'[0]} \cdot x'[i-l]$$

This is

$$a[k] = -\frac{a'[k]}{a'[0]} = \frac{f_{max}[mx-k]}{f_{max}[mx]} \quad \text{for } k=1 \text{ to } mx$$

and

$$b[l] = \frac{b'[l]}{a'[0]} = \frac{f_{max}[l]}{f_{max}[mx]} \quad \text{for } l=0 \text{ to } mx$$

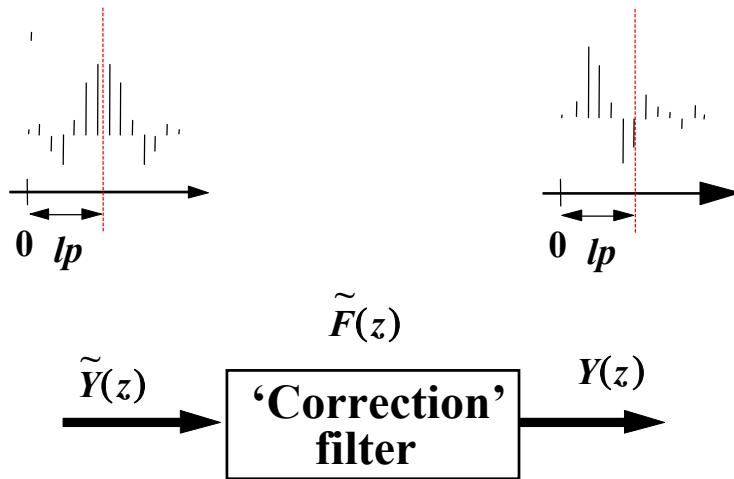
The convolution sum

$$y'[i] = \sum_{k=1}^{mx} a[k] \cdot y'[i-k] + \sum_{l=0}^{mx} b[l] \cdot x'[i-l]$$

$y'[i]$ = the time reversed ‘corrected’ sequence.

To obtain $y[i]$ flip $y'[i]$ back in time.

'Corrected' seismogram: $Y(z) = \tilde{F}(z) \cdot \tilde{Y}(z)$



$\tilde{F}(z) =$ changes the linear phase filter $F(z)$ into a minimum phase filter

\Rightarrow onset of output signal is advanced by lp samples.

To account for this time shift in the corrected seismogram:

- a) change time tag of ‘corrected’ trace

or

- b) delay ‘corrected’ trace by lp samples.

What is needed for ‘correction’?

$mx + 1$ coefficients of the maximum phase portion of the a linear phase FIR filter

How to get maximum phase component?

$$F(z) = \sum_{l=0}^m b_l z^{-l} = b_0 \prod_{l=1}^m (1 - c_l z^{-l})$$

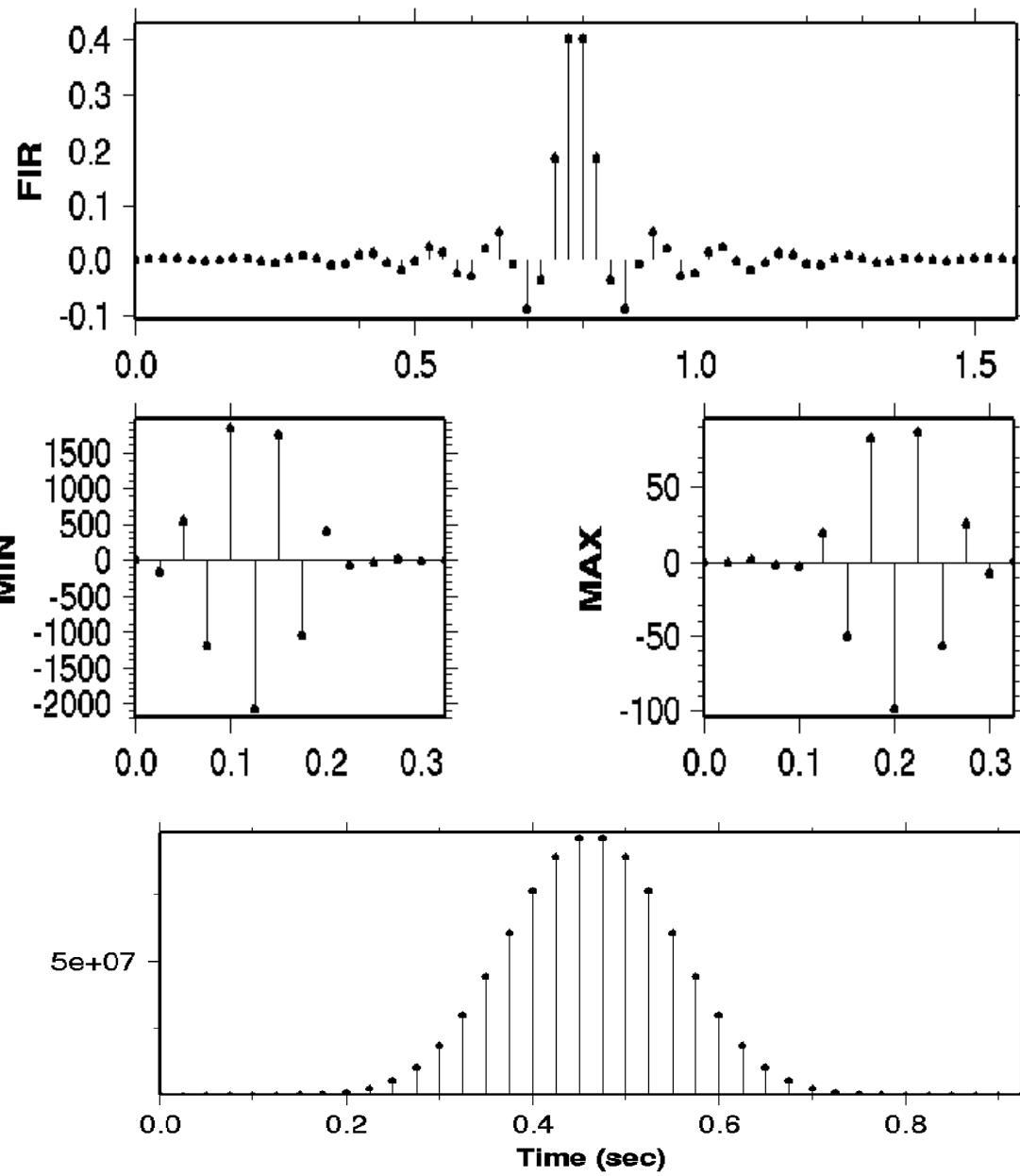
mn zeros inside UC (c_i^{min}) $\Leftrightarrow F_{min}(z)$

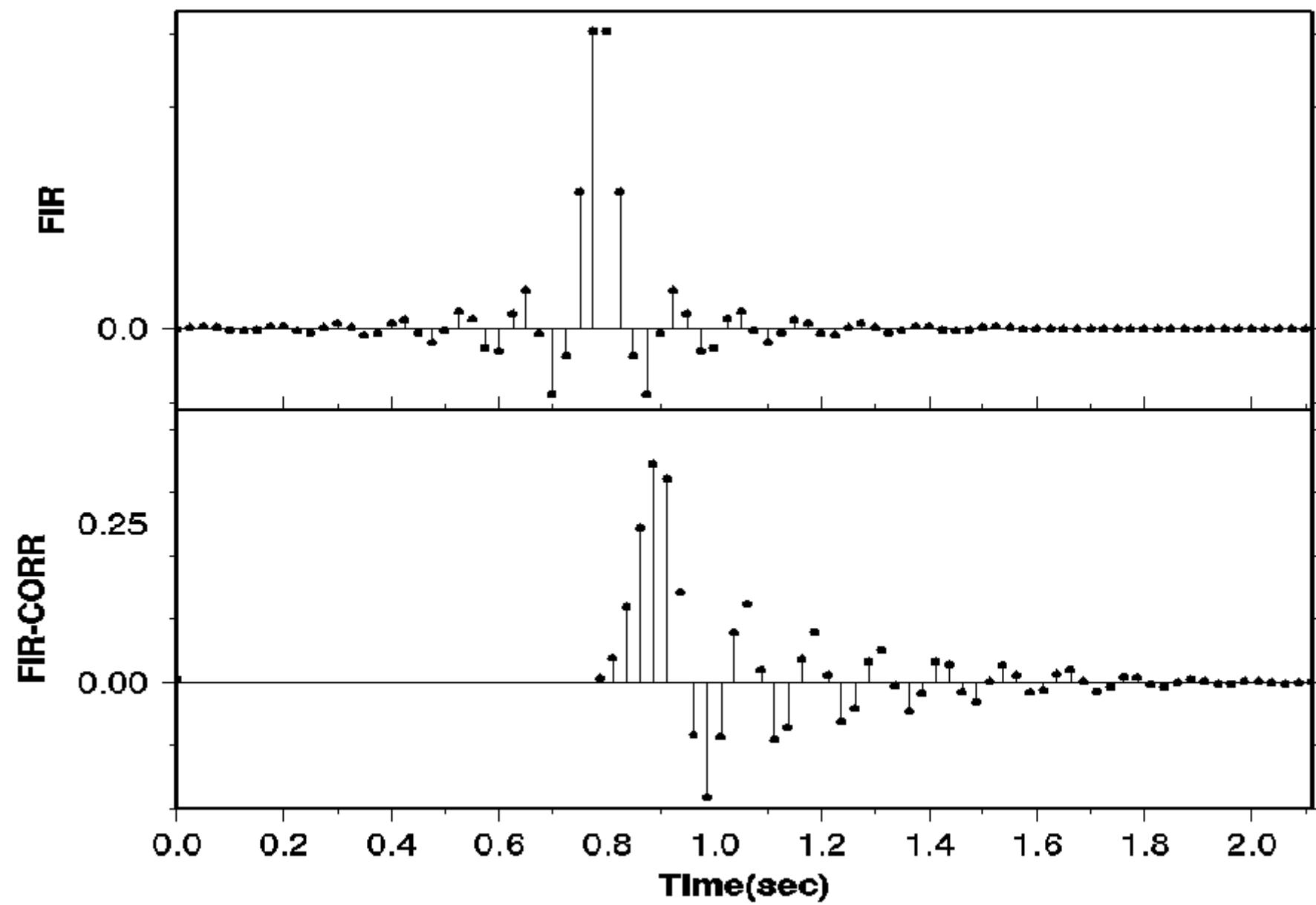
mx zeros outside UC (c_i^{max}) $\Leftrightarrow F_{max}(z)$

$$F_{max}(z) = \sum_{i=0}^{mx} f_i^{max} z^{-i} = b_0 \prod_{i=1}^{mx} (1 - c_i^{max} z^{-i})$$

From zeros outside UC \rightarrow polynomial $F_{max}(z)$.

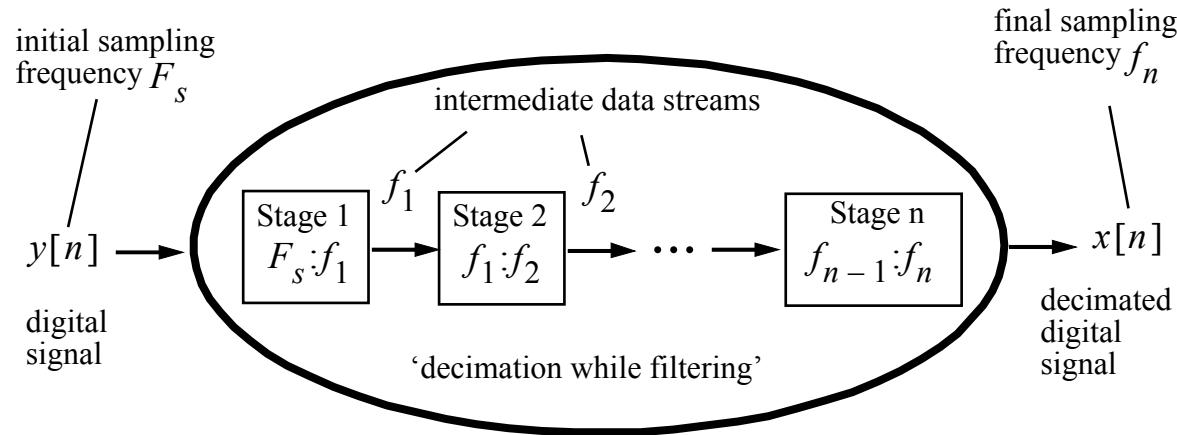
$f_{max}[l]$ for $l = 0$ to mx : coefficients of polynomial $F_{max}(z)$.





Correction Procedure

Decimation stages



Full Correction (not always necessary):

interpolation $\rightarrow f_{n-1}$ \rightarrow correction for FIR filter stage n

interpolation $\rightarrow f_{n-2}$ \rightarrow correction for FIR filter stage n-1

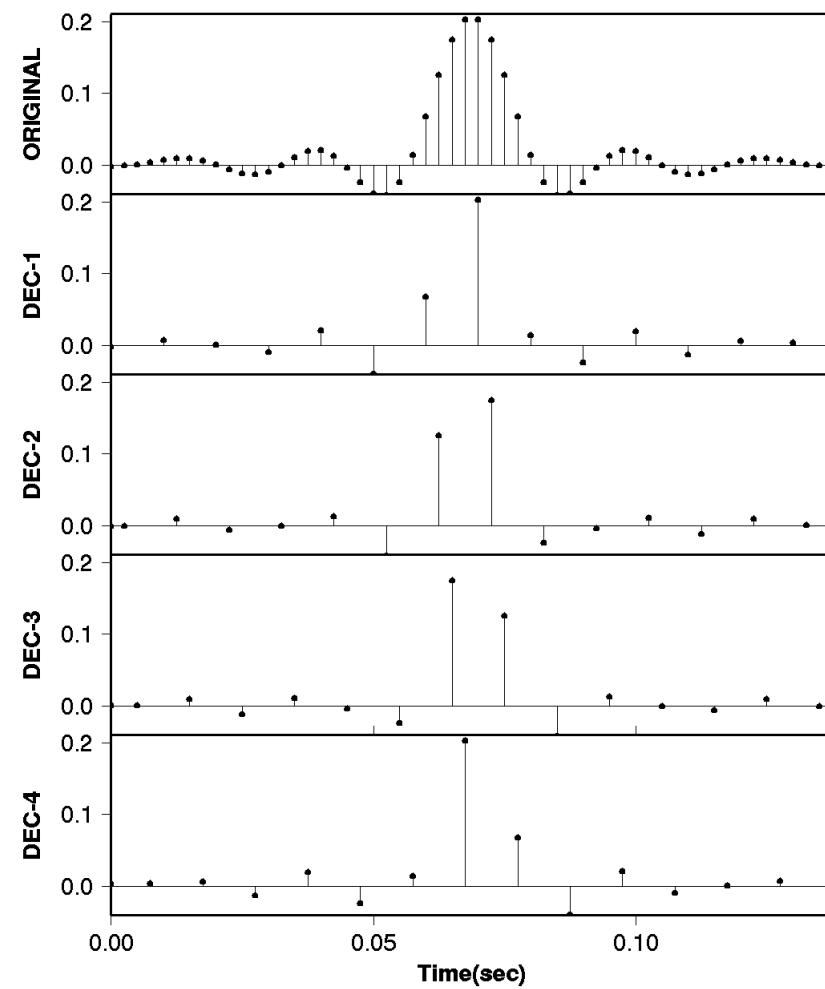
⋮

interpolation $\rightarrow f_1$ \rightarrow correction for FIR filter stage 2

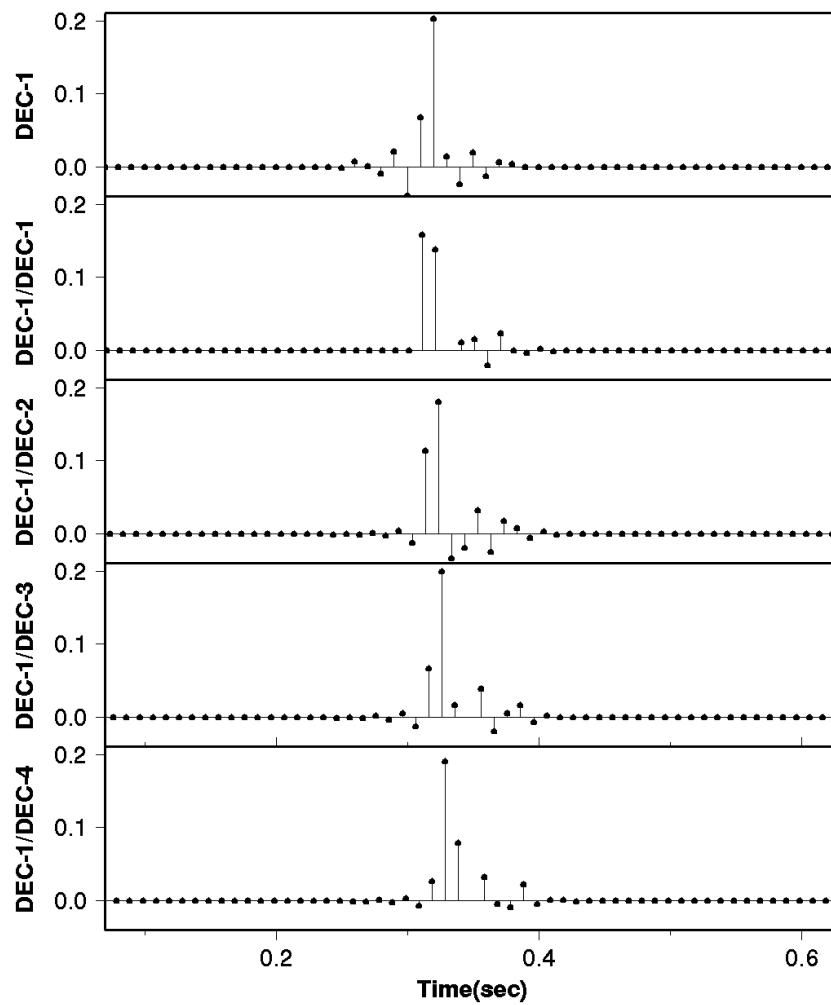
interpolation $\rightarrow F_s$ \rightarrow correction for FIR filter stage 1

finally: decimation back to f_n

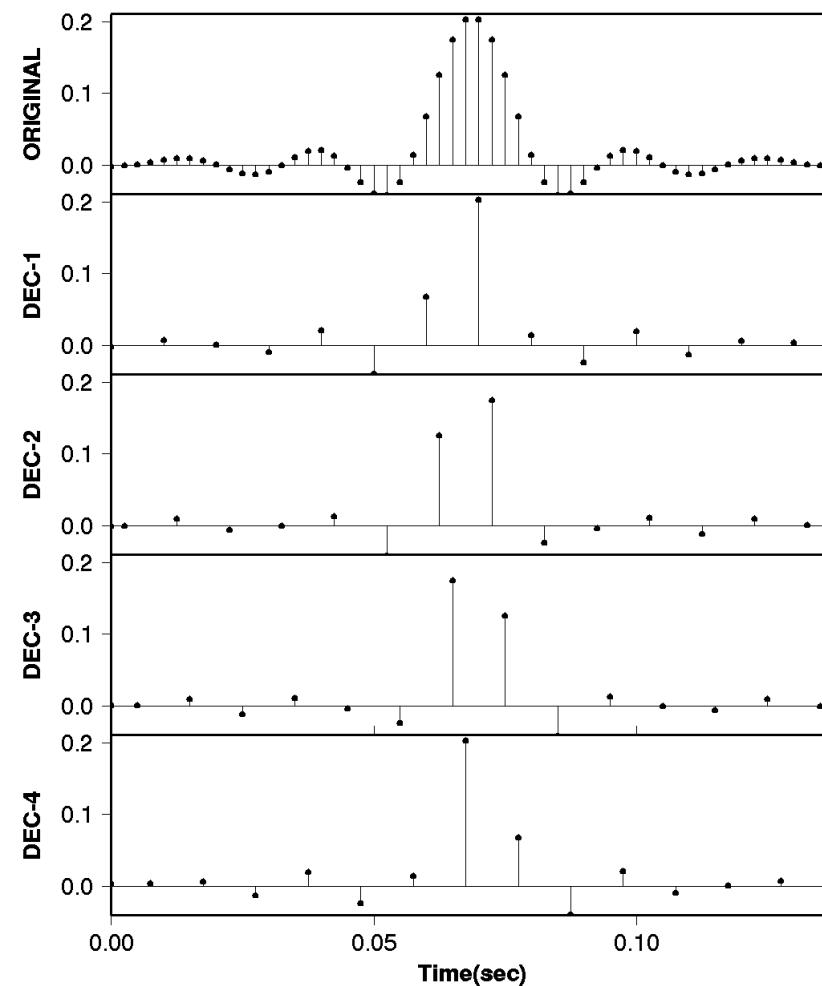
SIL SYSTEM 400/100 Hz



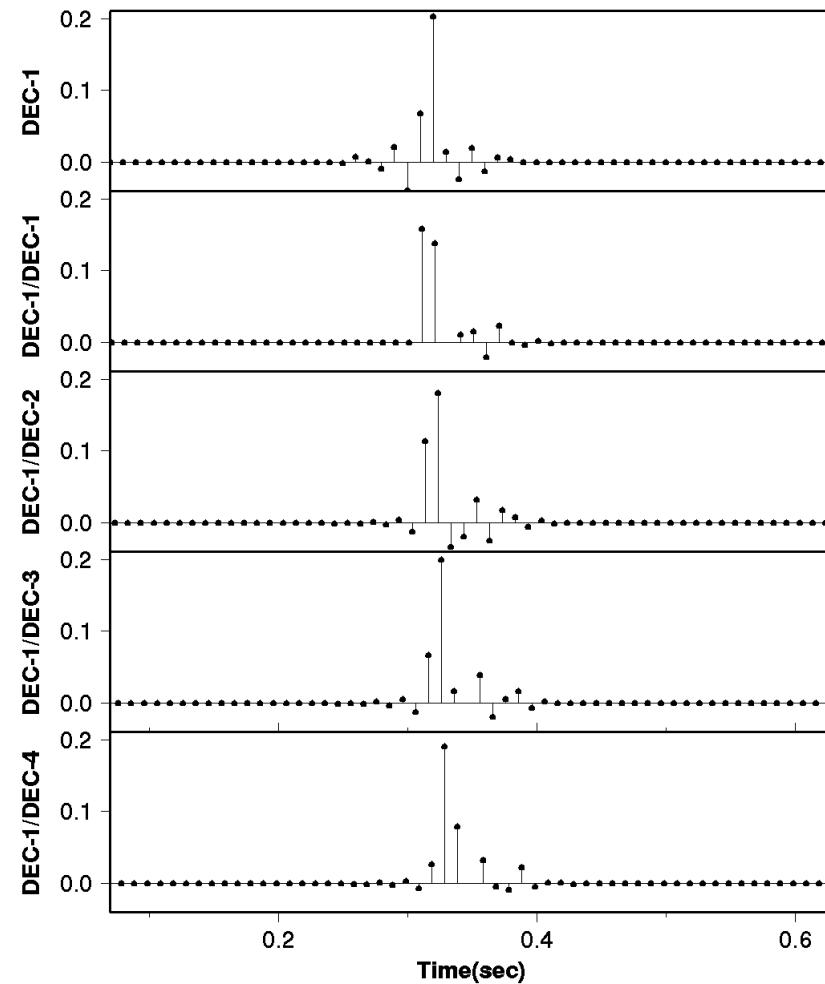
SIL FIR CORRECTION (100 Hz)



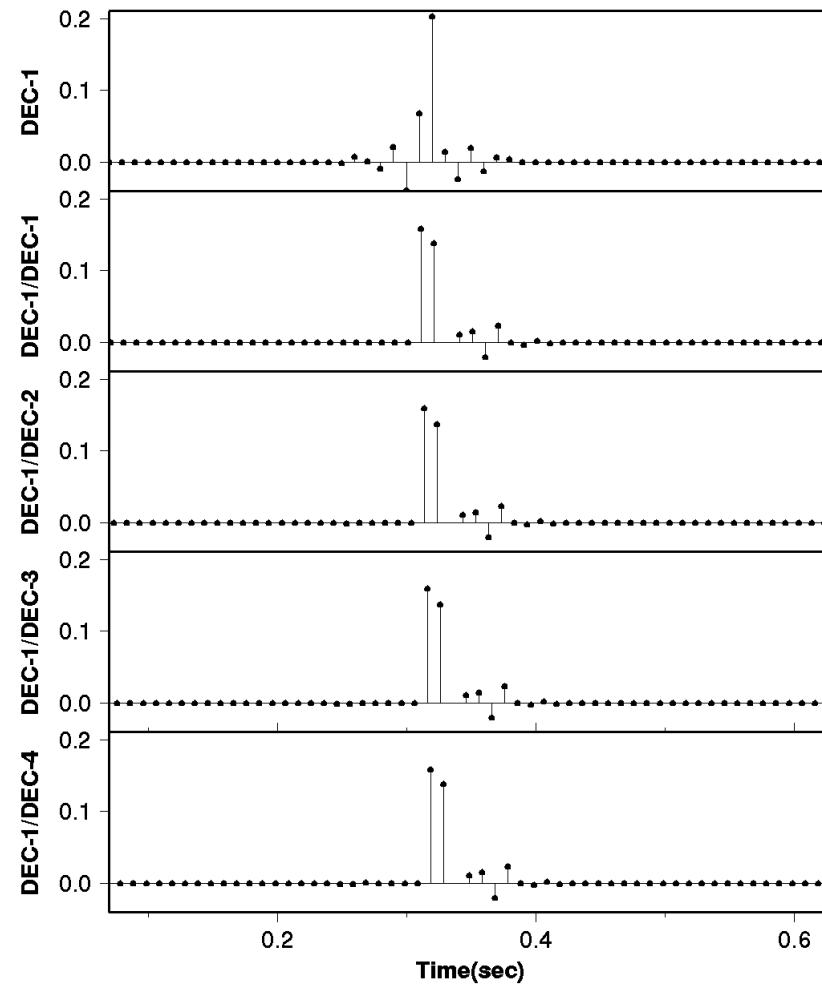
SIL SYSTEM 400/100 Hz

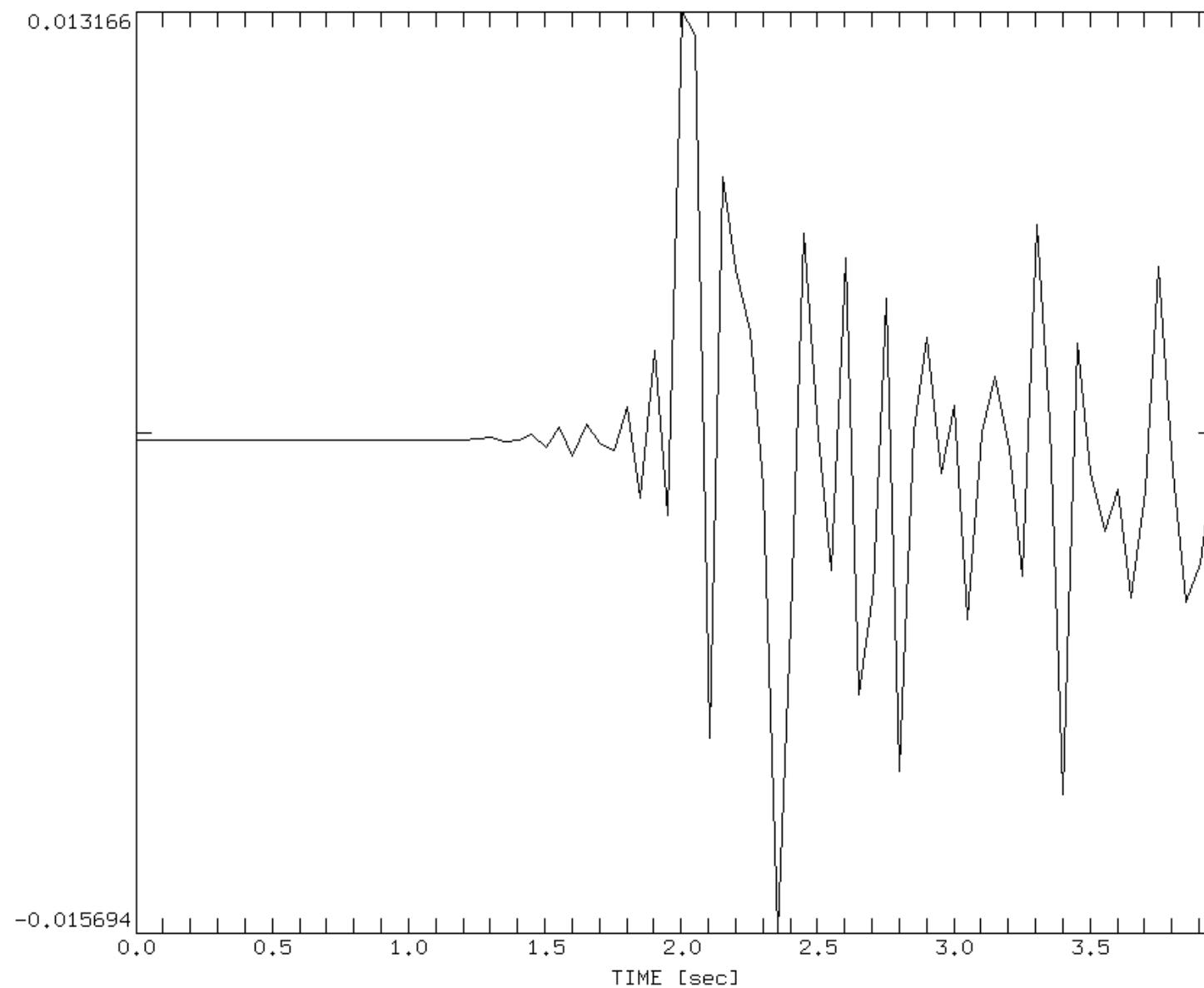


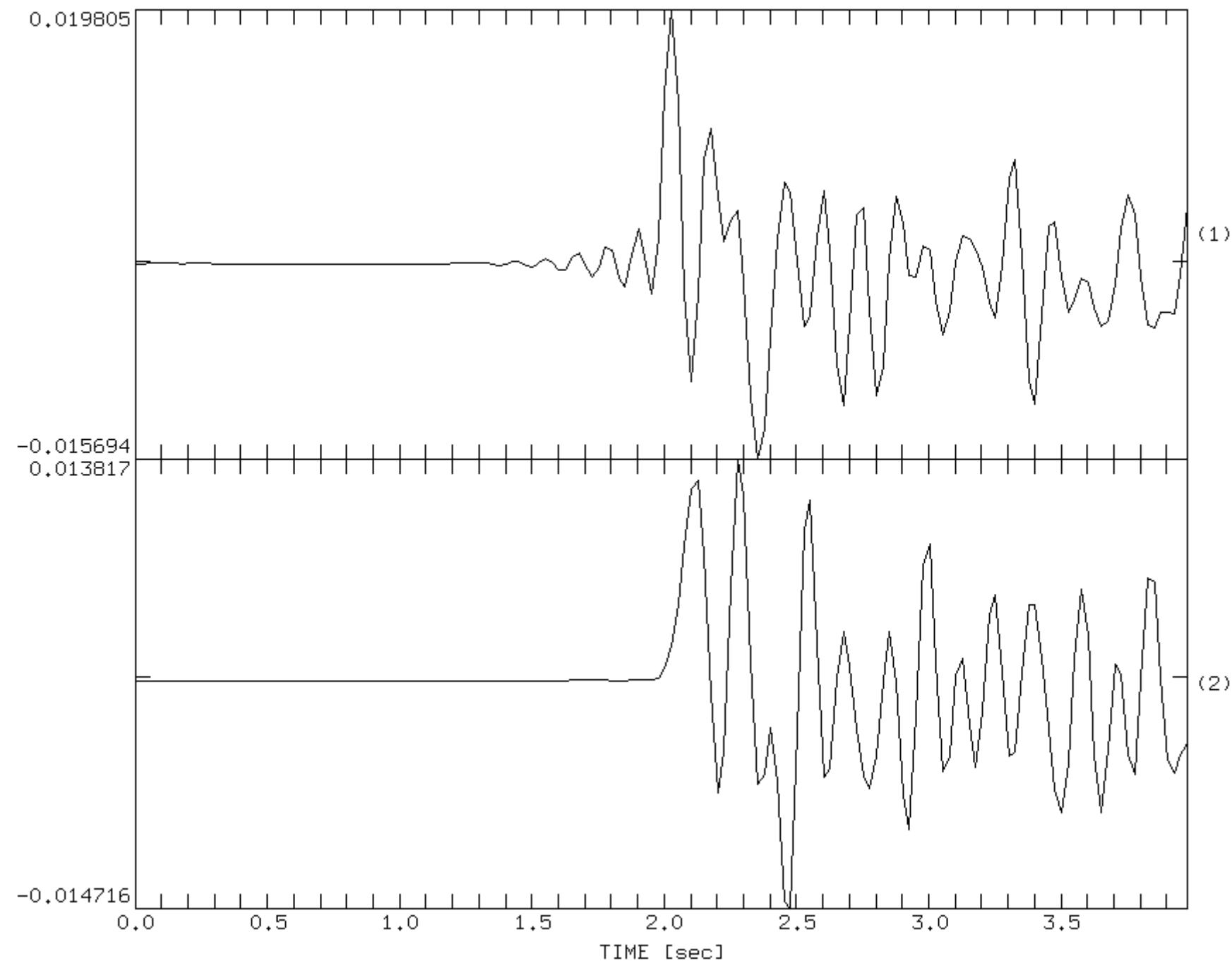
SIL FIR CORRECTION (100 Hz)

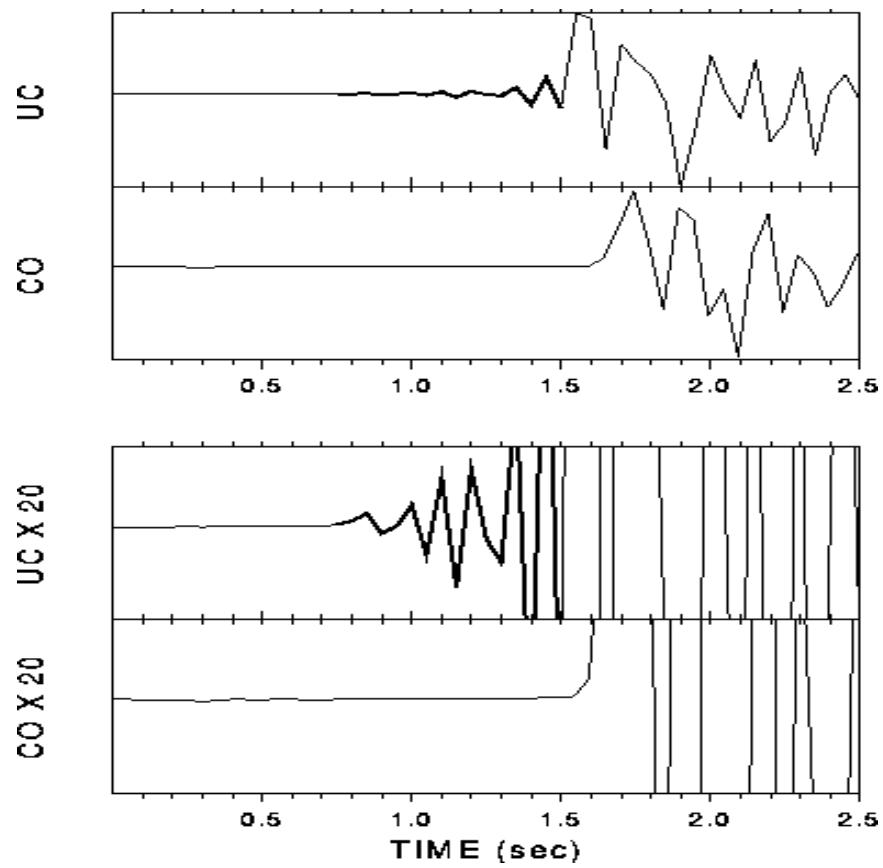


SIL FIR PLUS LINEAR PHASE CORRECTION (100 Hz)

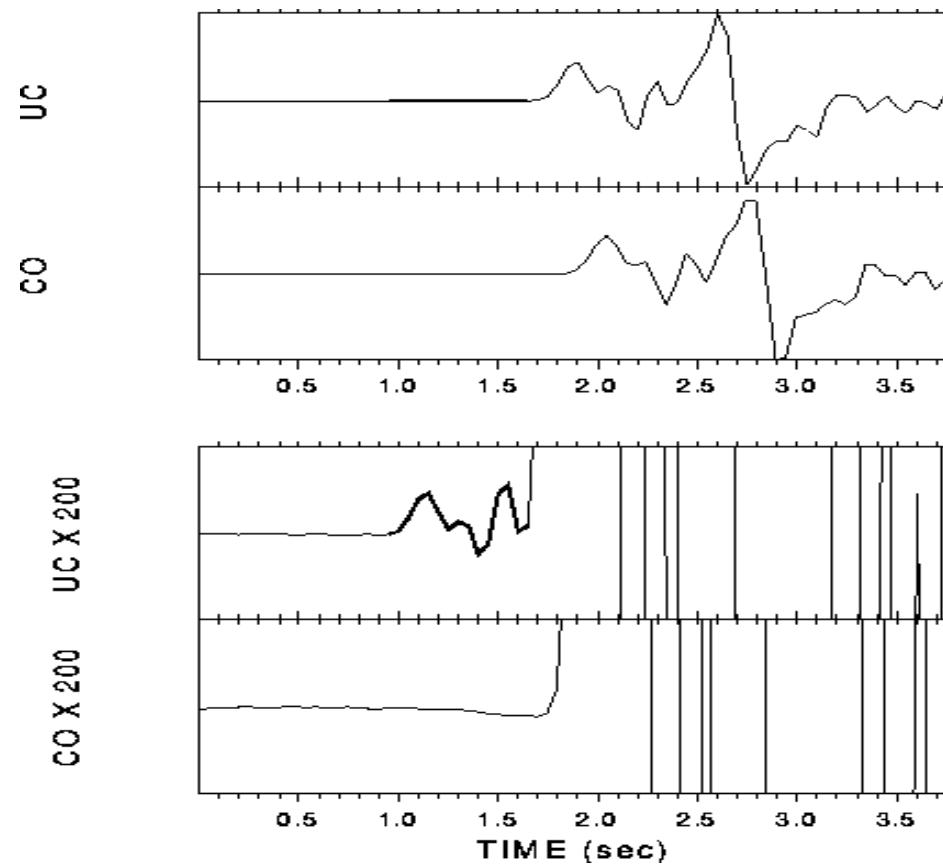




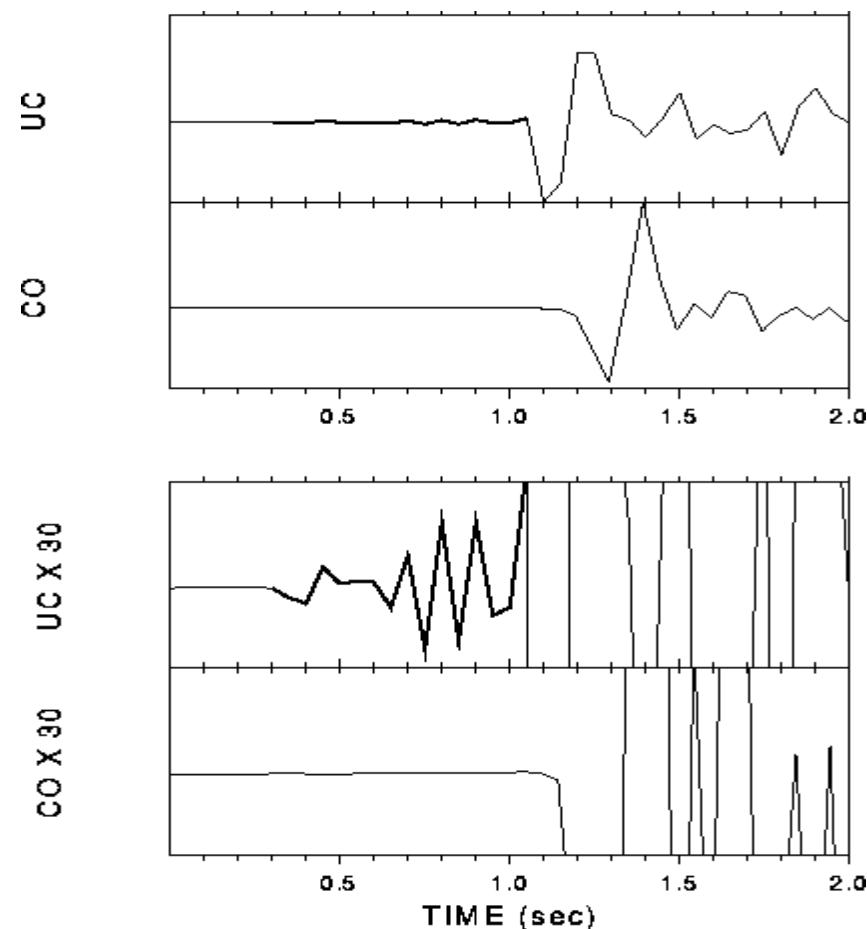




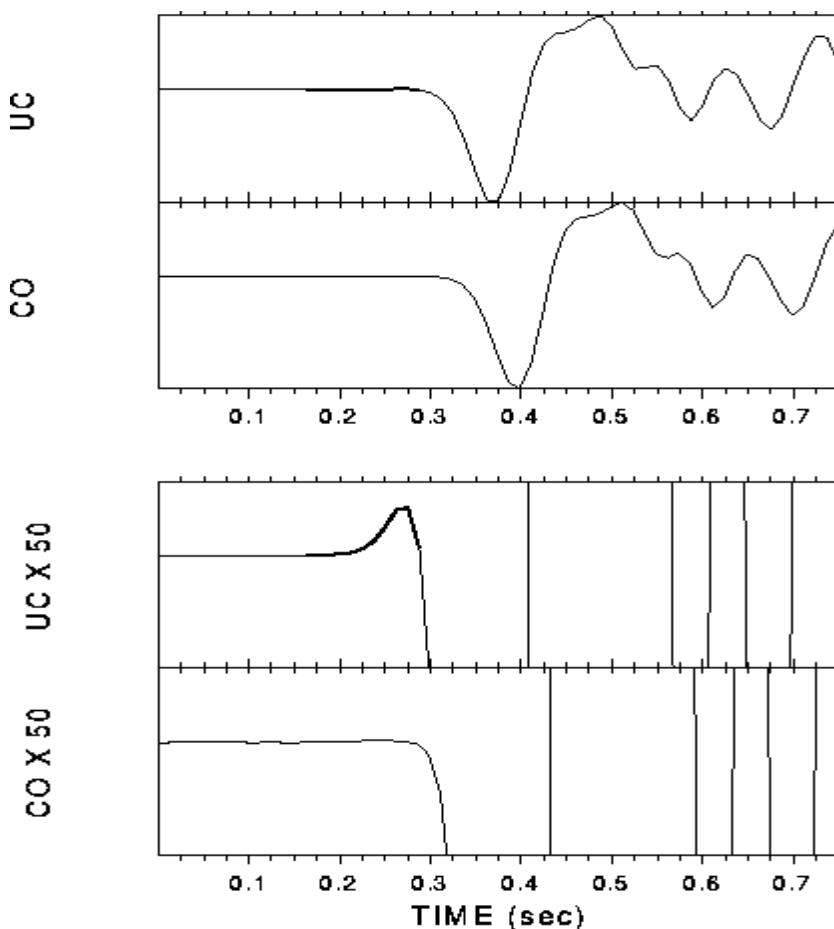
a)



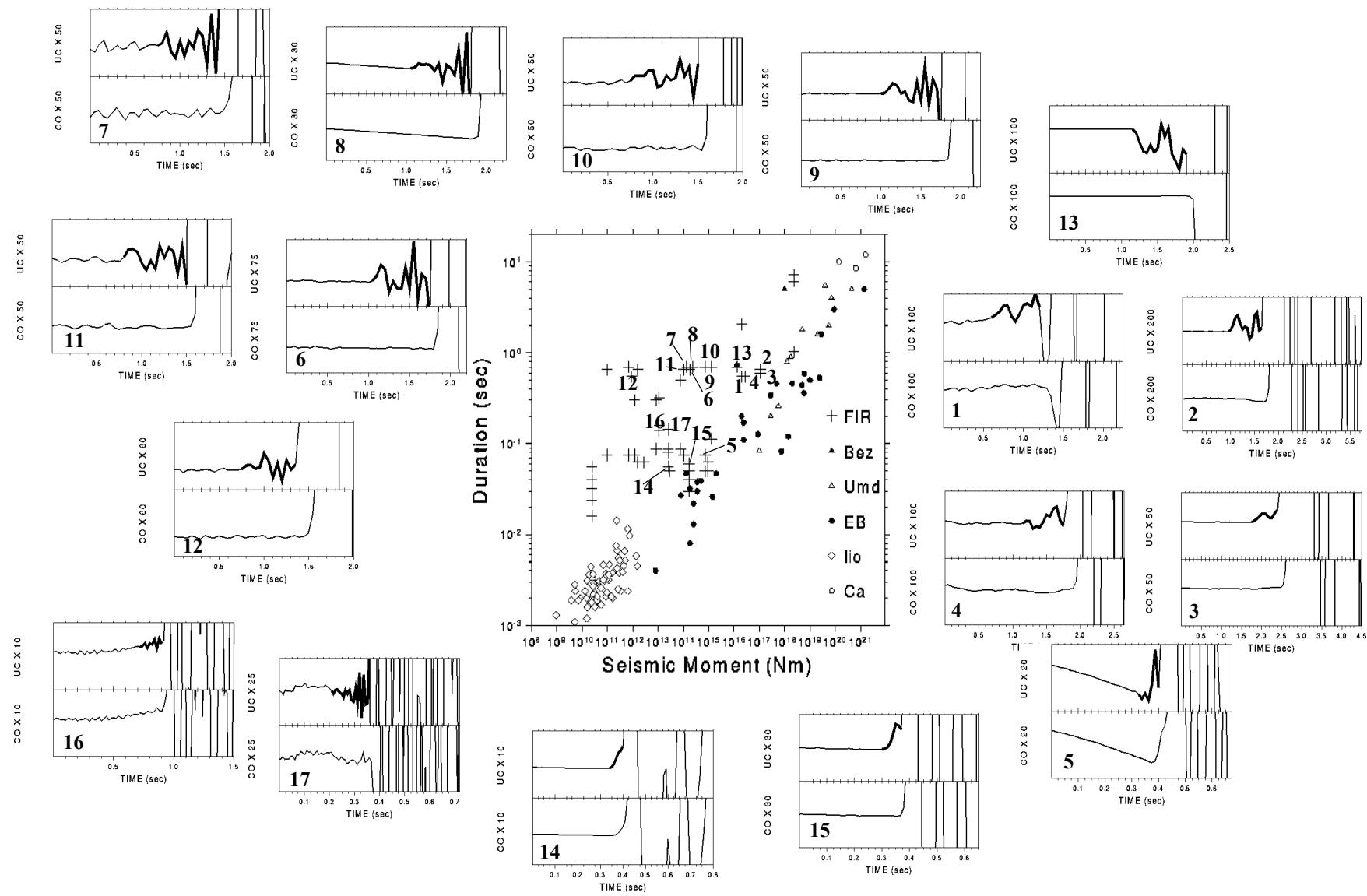
b)



a)



b)



Conclusions

FIR filter generated precursory artefacts:

- can become impossible to be identified visually
- can have similar scaling properties as nucleation phases

Zero - phase FIR filters in general

- affect the determination of all onset properties (onset times, onset polarities)

Consequence

For the interpretation of onset properties (onset times, onset polarities, nucleation phases, etc.) the acausal response of the zero-phase FIR filter has to be removed

but not

for waveform analysis.